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(54) Data modulation method, and related modulator and demodulator devices

(57) In a single-carrier or multi-carrier modulator (MOD), data symbols are modulated on at least one carrier in accordance with at least one constellation scheme. Idle data symbols are replaced by null data symbols and the latter null data symbols are associated with a low power constellation point of the at least one constellation scheme. The constellation points associ-

ated with the data symbols are randomly transformed into constellation points of equal power. In this way, idle data symbols are transferred with minimised power and sufficiently randomised to allow equaliser training and/or tracking at the receiving demodulator (DMOD) and to avoid high crest factors in multi-carrier systems.

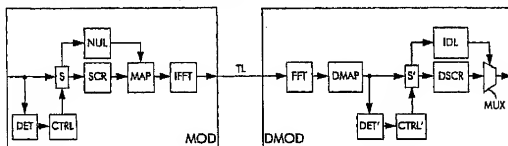


Fig. 1

Description

[0001] The present invention relates to a modulation method as defined in the non-characteristic part of claim 1, to a modulator device as defined in the non-characteristic part of claim 7, and to a demodulator device as defined in the non-characteristic part of claim 9.

[0002] Such a modulation method and devices to perform respectively the modulation and demodulation are already known in the art, e.g. from the Nortel standard contribution entitled 'On the ATM TG for VDSL' from the author L.D. Humphrey. This standard contribution with reference 'TD44 Antwerp' was presented on the meeting of 20-24th April 1998 in Antwerp of ETSI's Sub-Technical Committee TM6. Therein, a VDSL (Very High Bit Rate Digital Subscriber Line) modem is described which transfers ATM (Asynchronous Transfer Mode) cells over a VDSL (Very High Speed Digital Subscriber Line) network segment. If idle or unused ATM cells are supplied to the modem, the modem replaces these idle ATM cells with null ATM cells, i.e. cells whose header and payload sections are completely filled with zero bytes. The modem further associates with the null data symbols comprising the null ATM cells the origin point of a constellation diagram used to modulate the carrier(s). In this way, transmitted power is minimised during transfer of idle ATM cells and FEXT (Far End Crosstalk) is reduced significantly. If however a sequence of idle ATM cells is transferred as a sequence of origin points of a constellation diagram, detection at the demodulator becomes difficult because of the powerless property of the origin point of the constellation diagram. In the known method, eight subsequent carriers are summed to perform the detection. Moreover, equaliser training and tracking at the demodulator is hampered if a sequence of powerless data symbols arrives there, as a consequence of which the known modulation technique is not suitable for use in modems equipped with an adaptive equaliser whose taps are adjusted on the basis of information in the received data symbols, such as a DMT (Discrete Multi Tone) modem used in an ADSL (Asymmetric Digital Subscriber Line) environment.

[0003] It is an object of the present invention to provide a modulation method, and related modulator and demodulator devices of the above known type, but wherein detection of idle data symbols is simplified and wherein transfer of idle data symbols does not have disadvantageous effects on equaliser training and tracking.

[0004] According to the invention, this object is realised by the modulation method defined in claim 1, the modulator device defined in claim 7 and the demodulator device defined in claim 9.

[0005] Indeed, by associating a low power constellation point with idle cells, the transmitted power is minimised during idle data transmission but no zero power points are transmitted so that it is no longer necessary

to sum eight successive carriers to enable detection. In addition, by randomly transforming constellation points into constellation points of equal power, equaliser training and tracking is simplified whilst not increasing the transmitted power. Moreover, it is avoided that all idle data symbols are represented by in-phase carriers which would otherwise result in a high crest factor in multi-carrier systems.

[0006] It is to be noticed that the term 'comprising', used in the claims, should not be interpreted as being limitative to the means listed thereafter. Thus, the scope of the expression 'a device comprising means A and B' should not be limited to devices consisting only of components A and B. It means that with respect to the present invention, the only relevant components of the device are A and B.

[0007] Similarly, it is to be noticed that the term 'coupled', also used in the claims, should not be interpreted as being limitative to direct connections only. Thus, the scope of the expression 'a device A coupled to a device B' should not be limited to devices or systems wherein an output of device A is directly connected to an input of device B. It means that there exists a path between an output of A and an input of B which may be a path including other devices or means.

[0008] An additional feature of the present invention is defined in claim 2.

[0009] Thus, a first embodiment of the invention is realised by enabling a modulator to inverse randomly the real and imaginary pads of constellation points associated with data symbols. The constellation points then are mirrored vis-à-vis the origin point of the constellation scheme which does not change their power.

[0010] Another additional feature of the present invention is defined in claim 3.

[0011] Thus, a second embodiment of the invention is realised by enabling a modulator to inverse randomly the real or imaginary pads of constellation points associated with data symbols. The constellation points then are mirrored vis-à-vis the horizontal or vertical axis of the constellation scheme which does not change their power.

[0012] Yet another additional feature of the present invention is defined in claim 4.

[0013] Thus, a third embodiment of the present invention is realised by enabling a modulator to randomly change the phase of constellation points. Such a phase rotation also does not change the power of the constellation points. In particular, the phase may be randomly changed over integer multiples of 90 degrees as defined in claim 5.

[0014] A further feature of the modulation method according to the present invention is defined in claim 6.

[0015] Indeed, the present invention is very suitable for transport of ATM (Asynchronous Transfer Mode) cells over an ADSL (Asymmetric Digital Subscriber Line) or VDSL (Very High Bit Rate Digital Subscriber Line) network segment because power consumption is

critical therein and most of the time an ATM link is expected to transport idle ATM cells. Moreover, an ADSL or VDSL DMT receiver includes an equaliser that has to be trained and, in case of a multi-carrier ADSL/VDSL system, clipping due to a high crest factor may occur. The present invention adds to solve all of these problems.

[0016] As is indicated in claims 8 and 10, scrambling and descrambling idle data symbols can be deactivated.

[0017] The above mentioned and other objects and features of the invention will become more apparent and the invention itself will be best understood by referring to the following description of an embodiment taken in conjunction with the accompanying drawings wherein:

Fig. 1 represents a block scheme of embodiments of a modulator device MOD and demodulator device DMOD according to the present invention;

Fig. 2 represents a constellation diagram of a square modulator wherein the present invention is applied; and

Fig. 3 represents a constellation diagram of a linear modulator wherein the present invention is applied.

[0018] In Fig. 1 a modulator MOD is coupled via a telephone line TL to a demodulator DMOD. The telephone line TL may support one-way communication, i.e. data traffic from the modulator MOD to the demodulator DMOD, or it may be a bi-directional link allowing communication in both directions. In the latter situation, the modulator MOD and demodulator DMOD form part of two communicating modems which each include the functionality of both the modulator MOD and the demodulator DMOD. This is for instance so in a VDSL (Very High Speed Digital Subscriber Line) system.

[0019] The modulator MOD includes a switch S, a null cell generator NUL, a scrambler SCR, a mapper MAP, an inverse fast fourier transformer IFFT, an idle cell detector DET and a control unit CTRL. The switch S, the scrambler SCR, the mapper MAP and the inverse fast fourier transformer IFFT are cascade connected between an input and output of the modulator MOD. The null cell generator NUL, coupled between an output of the switch S and input of the mapper MAP, constitutes a bypass for the scrambler SCR. The input of the modulator MOD is interconnected with an input of the idle cell detector DET and the idle cell detector DET further has an output coupled to an input of the control unit CTRL. The control unit CTRL has an output connected to a control input of the switch S.

[0020] The demodulator DMOD includes a fast fourier transformer FFT, a demapper DMAP, a switch S', a descrambler DSCR, an idle cell generator IDL, a multiplexer MUX, a null data detector DET' and a control unit CTRL'. The fast fourier transformer FFT, the demapper DMAP, the switch S', the descrambler DSCR and the multiplexer MUX are cascade connected between an input and output of the demodulator DMOD. The idle

cell generator IDL, coupled between a second output of the switch S' and second input of the multiplexer MUX, constitutes a bypass for the descrambler DSCR. The null data detector DET' has an input interconnected with an output of the demapper DMAP and an output connected to an input of the control unit CTRL'. The control unit CTRL' is provided with an output connected to a control input of the switch S'.

[0021] In the following paragraphs, the working of the modulator MOD and demodulator DMOD of Fig. 1 will be described in case non-idle data are to be transferred there between and in case idle data are to be transferred there between. It is assumed that the modulator MOD is a DMT (Discrete Multi Tone) modulator, that the demodulator DMOD is a DMT demodulator, and that both are used in a VDSL (Very High Speed Digital Subscriber Line) system. The system drawn in Fig. 1 in other words is a multi-carrier system wherein different amounts of bits are modulated on different carriers in accordance with different constellation schemes. The amount of bits modulated on each carrier may for instance depend on the noise measured on the telephone line TL at the frequency of that carrier. In fact however, it is not important in view of the present invention how the amount of bits modulated on a carrier is determined.

[0022] At the input of the modulator MOD, ATM (Asynchronous Transfer Mode) cells arrive that have to be transferred over the VDSL (Very High Speed Digital Subscriber Line) network segment between the modulator MOD and demodulator DMOD. If non-idle ATM cells arrive at the entrance of the modulator MOD, the scrambler SCR adds a random sequence of bits to the bits of these non-idle ATM cells to obtain a good statistic of the transmuted signal. The so scrambled bits are then applied to the mapper MAP which allocates to each carrier of a DMT (Discrete Multi Tone) symbol a data symbol comprising the amount of bits that has to be modulated on this carrier in accordance with a bit allocation algorithm. In addition, the phase and amplitude of the carriers are modulated by the mapper MAP in accordance with an appropriate constellation scheme. The mapper MAP thereto associates with each data symbol a constellation point of the constellation scheme and then transforms this constellation point into another constellation point having the same power. A carrier wherein four bits, B0, B1, B2 and B3, are to be modulated, is for instance modulated in accordance with the constellation scheme drawn in Fig. 2. Instead of the constellation point associated with a data symbol by the scheme of Fig. 2, the carrier is modulated with a constellation point obtained by randomly rotating the former point over an integer multiple of 90 degrees. A carrier wherein only two bits, B0 and B1, have to be modulated, is modulated in accordance with the constellation scheme of Fig. 3. Instead of the constellation point associated with a data symbol by the scheme in Fig. 3, the carrier is modulated with a constellation point

obtained by randomly mirroring the former point vis-à-vis the origin point in Fig. 3. Once the phases and amplitudes of the carriers are modulated, the set of carriers constituting one DMT (Discrete Multi Tone) symbol is transformed from frequency domain to time domain by the inverse fast fourier transformer IFFT. The time domain DMT (Discrete Multi Tone) symbol is transferred over the telephone line TL to the demodulator DMOD wherein the fast fourier transformer FFT transforms the time domain DMT (Discrete Multi Tone) symbol again into a frequency domain DMT (Discrete Multi Tone) symbol. The demapper DMAP uses the same constellation schemes as the mapper MAP in the modulator MOD and de-randomises the assignment of constellation points to subtract from each carrier the amount of bits modulated thereon and the descrambler DSCR subtracts the random sequence of bits added by the scrambler SCR so that at the output of the demodulator DMOD, an exact copy of the ATM cells which were applied to the input of the modulator MOD appears.

[0023] In case idle ATM cells are applied to the input of the modulator MOD, the idle cell detector DET recognises these idle ATM cells and deactivates the scrambler SCR via the control unit CTRL and switch S. The idle ATM cells in other words bypass the scrambler SCR via the null cell generator NUL which converts each idle ATM cell into a null cell, i.e. calls whose header and payload section is completely filled with zero bits. These zero bits are then applied to the mapper MAP to be modulated on the different carriers in accordance with the respective constellation schemes. If for instance a sequence of zero bits has to be modulated on a carrier with constellation scheme drawn in Fig. 2, the lowest power constellation point in the first quadrant of the constellation scheme should be transmitted until non-idle cells arrive at the modulator MOD. Since this may cause difficulties for equaliser training, the sequence of zero bits is converted by the mapper MAP into a random sequence of low power constellation points in the 4 quadrants of the constellation scheme. More particularly, the mapper MAP generates a random sequence of the four black filled constellation points in Fig. 2. The mapper MAP in other words again randomly rotates the constellation points associated with the data symbols over an integer multiple of 90 degrees. The average power to be transferred compared to a modem which encodes all null data symbols into the low power constellation point of the first quadrant is not increased. The inverse fast fourier transformer IFFT again transforms the modulated carriers constituting one DMT (Discrete Multi Tone) symbol into a time domain DMT (Discrete Multi Tone) symbol that is transferred over the telephone line TL towards the demodulator DMOD. In the demodulator DMOD, the fast fourier transformer FFT regenerates the frequency domain DMT (Discrete Multi Tone) symbols from the time domain DMT (Discrete Multi Tone) symbols. The demapper DMAP de-randomises the low power constellation points and sub-

tracts from each carrier the amount of zero bits modulated thereon. The null data detector DET recognises these zero bits and controls the switch S via the control unit CTRL so that the null bits are applied to the idle cell generator IDL which charges null ATM cells for idle ATM cells. The controller CTRL thereto controls switch S so that the descrambler DSCR is bypassed. The multiplexer MUX multiplexes idle and non-idle data into a single outgoing information stream.

[0024] In case of square modulation, the phase of this carrier is randomised as indicated by the four low power constellation points in the different quadrants of the constellation scheme in Fig. 2. In case of linear modulation (amplitude modulation without phase modulation) on the other hand, null data cells are converted into a random sequence of two inverse constellation points, as is illustrated by Fig. 3.

[0025] A first remark is that, although the multi-carrier signal in the above described embodiment is transported over a telephone line TL, the applicability of the present invention is not restricted by the transmission medium via which the signal is transported. In particular, transmission over any kind of connection between the modulator MOD and demodulator DMOD, e.g. a cable connection, a satellite connection, a radio link through the air, and so on, may be improved according to the present invention.

[0026] The invention also is not only related to ADSL (Asymmetric Digital Subscriber Line) or similar systems wherein DMT (Discrete Multi Tone) modulation is used. A person skilled in the art will be able to adapt the above described embodiment so that it is applicable in any other system wherein a single-carrier signal or multi-carrier signal is transmitted from a transmitting modem to a receiving modem.

[0027] Yet another remark is that embodiments of the present invention are described above in terms of functional blocks. From the functional description of these blocks, given above, it will be apparent for a person skilled in the art of designing electronic devices how embodiments of these blocks can be manufactured with well-known electronic components. A detailed architecture of the contents of the functional blocks hence is not given.

[0028] While the principles of the invention have been described above in connection with specific apparatus, it is to be clearly understood that this description is made only by way of example and not as a limitation on the scope of the invention.

Claims

1. Modulation method for modulating a sequence of data symbols on at least one carrier in accordance with at least one constellation diagram whereby idle data symbols in said sequence are replaced by null data symbols before said at least one carrier is modulated.

CHARACTERISED IN THAT a low power constellation point is associated with said null data symbols and constellation points associated with said data symbols are randomly transformed into constellation points of equal power.

2. Modulation method according to claim 1,
CHARACTERISED IN THAT real and Imaginary
parts of said constellation points are randomly
inverted whereby said constellation points are
transformed into constellation points of equal
power.
3. Modulation method according to claim 1,
CHARACTERISED IN THAT real or Imaginary
parts of said constellation points are randomly
inverted whereby said constellation points are
transformed into constellation points of equal
power.
4. Modulation method according to claim 1,
CHARACTERISED IN THAT said constellation
points are rotated in phase to be transformed into
constellation points of equal power.
5. Modulation method according to claim 4
CHARACTERISED IN THAT said phase is rotated
over an integer multiple of 90 degrees.
6. Modulation method according to any of claims 1 to
5,
CHARACTERISED IN THAT said sequence of data
symbols comprises an Asynchronous Transfer
Mode (ATM) cell sequence, said idle data symbols
comprise idle Asynchronous Transfer Mode (ATM)
cells and said at least one carrier is modulated in
accordance with a Digital Subscriber Line (DSL)
physical layer protocol.
7. Modulator device (MOD) for modulating a
sequence of data symbols on at least one carrier in
accordance with at least one constellation diagram,
said modulator device (MOD) including means
(NUL) for replacing said idle data symbols in said
sequence by null data symbols,
CHARACTERISED IN THAT said modulator device
(MOD) further comprises means (MAP) to associ-
ate a low power constellation point with said null
data symbols and to randomly transform constella-
tion points associated with said data symbols into
constellation points of equal power of said at least
one constellation diagram for modulation of said at
least one carrier.
8. Modulator device (MOD) according to claim 7,
CHARACTERISED IN THAT said modulator device
(MOD) further comprises detection means (DET) to
detect said idle data symbols and scrambling

means (SCR) for scrambling said data symbols,
said scrambling means (SCR) being adapted to be
deactivated if said detection means (DET) detects
said idle data symbols.

9. Demodulator device (DMOD) for demodulating a
sequence of data symbols from at least one carrier
in accordance with at least one constellation dia-
gram, said demodulator device (DMOD) including
means (IDL) to replace null data symbols demodu-
lated from said at least one carrier by idle data sym-
bols,
CHARACTERISED IN THAT said demodulator
device (DMOD) is adapted to cooperate with a
modulator device (MOD) as defined in claim 6 and
thereto comprises means to de-randomise constella-
tion points associated with said data symbols and
to associate with a low power constellation point
said null data symbols.
10. Demodulator device (DMOD) according to claim 9,
CHARACTERISED IN THAT said demodulator
device (DMOD) further comprises detection means
(DET) to detect said null data symbols and
descrambling means (DSCR) for descrambling said
data symbols, said descrambling means (DSCR)
being adapted to be deactivated if said detection
means (DET) detects said null data symbols.

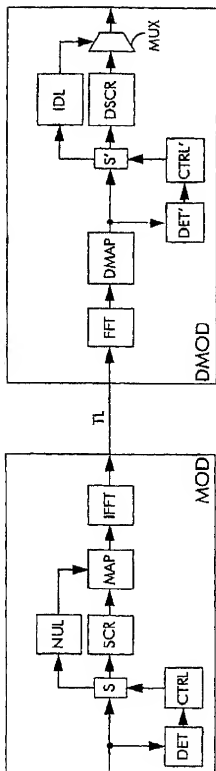


Fig. 1

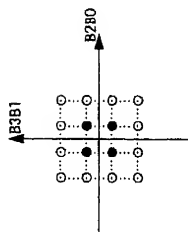


Fig. 2

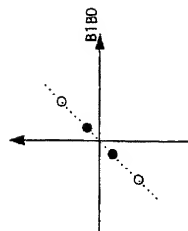


Fig. 3

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EUROPEAN SEARCH REPORT

Application Number
EP 98 40 1298

| DOCUMENTS CONSIDERED TO BE RELEVANT | | |
|---|--|--|
| Category | Citation of document with indication, where appropriate, of relevant passages | Relevant to claim |
| A | US 5 541 964 A (WEITZNER ANDREW T ET AL) 30 July 1996 * column 2, line 1 - line 19 * * column 4, line 50 - column 5, line 6 * | 1,7 |
| A | WO 96 42160 A (PHILIPS ELECTRONICS NV :PHILIPS NORDEN AB (SE)) 27 December 1996 * claims 1,3,4 * | 1,7 |
| A | EP 0 808 074 A (MATSUSHITA ELECTRIC IND CO LTD) 19 November 1997 * column 2, line 49 - column 3, line 45 * | 1-10 |
| | | TECHNICAL FIELDS SEARCHED (IPC CLASS) |
| | | H04Q H04L |
| The present search report has been drawn up for all claims | | |
| Place of search | Date of completion of the search | Searcher |
| THE HAGUE | 10 November 1998 | Gregori, S |
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